### ON Semiconductor®



### BelaSigna 300

### Audio Processor for Portable Communication Devices

### 1.0 General Overview

#### 1.1 Introduction

BelaSigna 300 is a DSP-based mixed-signal audio processing system that delivers superior audio clarity without compromising size or battery life. The processor is specifically designed for monaural portable communication devices requiring high performance audio processing capabilities and programming flexibility when form-factor and power consumption are key design constraints.

The efficient dual-MAC 24-bit CFX DSP core, together with the HEAR configurable accelerator signal processing engine, high speed debugging interface, advanced algorithm security system, state-of-the-art analog front end, Class D output stage and much more, constitute an entire system on a single chip, which enables manufacturers to create a range of advanced and unique products. The system features a high level of instructional parallelism, providing highly efficient computing capability. It can simultaneously execute multiple advanced adaptive noise reduction and echo cancellation algorithms, and uses an asymmetric dual-core patented architecture to allow for more processing in fewer clock cycles, resulting in reduced power consumption.

BelaSigna 300 is supported by a comprehensive suite of development tools, hands-on training, full technical support and a network of solution partners offering software and engineering services to help speed product design and shorten time to market.

### 1.2 Key Features

- Flexible DSP-based system: a complete DSP-based, mixed-signal audio system consisting of the CFX core, a fully programmable, highly cycle-efficient, dual-Harvard architecture 24-bit DSP utilizing explicit parallelism; the HEAR configurable accelerator for optimized signal processing; and an efficient input/output controller (IOC) along with a full complement of peripherals and interfaces, which optimize the architecture for audio processing at extremely low power consumption
- Excellent audio fidelity: up to 110dB input dynamic range, exceptionally low system noise and low group delay
- Ultra-low-power: typically 1-5mA
- Miniature Form Factor: available in a miniature 3.63mm x 2.68mm x 0.92mm (including solder balls) WLCSP package
- Multiple audio input sources: four input channels from four input sources can be used simultaneously for multiple microphones or direct analog audio inputs
- Full range of configurable interfaces: including a fast I<sup>2</sup>C-based interface for download, debug and general communication, a highly configurable PCM interface to stream data into and out of the device, a high-speed UART, an SPI port and 5 GPIOs
- Integrated A/D converters and powered output: minimize need for external components
- Flexible clocking architecture: supports speeds up to 40MHz
- "Smart" power management: including low current standby mode requiring only 0.05mA
- Diverse memory architecture: 4864x48-bit words of shared memory between the CFX core and the HEAR accelerator plus
   8-Kword DSP core data memory, 12-Kwords of 32-bit DSP core program memory as well as other memory banks
- Data security: sensitive program data can be encrypted for storage in external NVRAM to prevent unauthorized parties from gaining access to proprietary software intellectual property, 128-bit AES encryption
- Development tools: interface hardware with USB support as well as a full IDE that can be used for every step of program development including testing and debugging

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### 2.0 Mechanical Information and Circuit Design Guidelines

### 2.1 Mechanical Information

BelaSigna 300 is available in an ultra-miniature wafer-level chip scale package (WLCSP) measuring only 3.63mm x 2.68mm.

The BelaSigna 300 WLCSP option is Green (RoHS-compliant). Contact ON Semiconductor for supporting documentation.

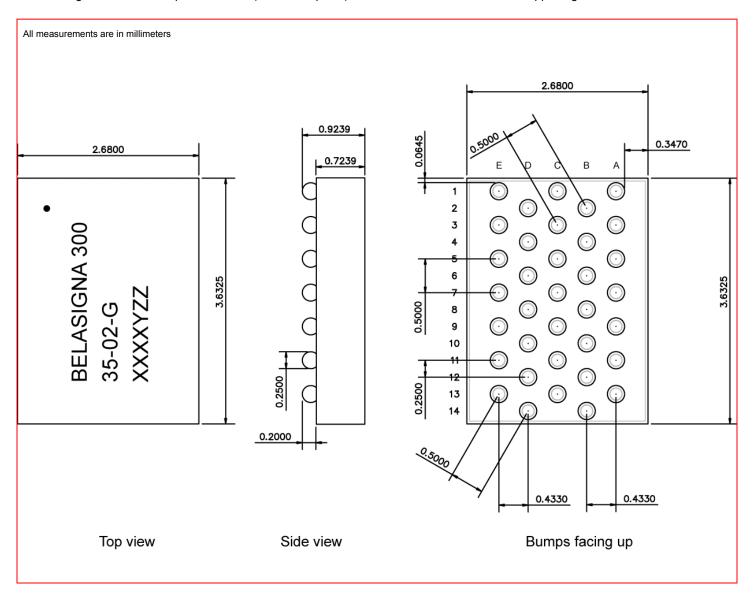


Figure 1: WLCSP Mechanical Information

A total of 35 active pins are present on the BelaSigna 300 WLCSP package. They are organized in a staggered array. A description of these pins is given in Table 1.

Table 1: Pad Descriptions

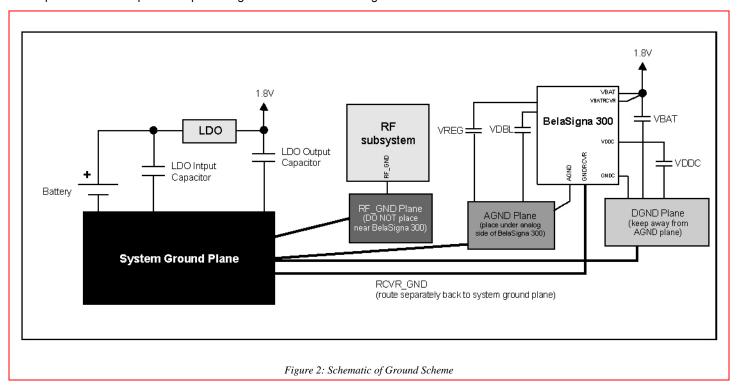
	BelaSigna 300 Pad Name	Description	I/O	A/D
A1	GNDRCVR	Ground for output driver	N/A	Α
A5	VBATRCVR	Power supply for output stage	I	Α
B2	RCVR_HP+	Extra output driver pad for high power mode	0	Α
C3	RCVR+	Output from output driver	0	Α
A3	RCVR-	Output from output driver	0	Α
B4	RCVR_HP-	Extra output driver pad for high power mode	0	Α
C5	CAP1	Charge pump capacitor pin 0	N/A	Α
B6	CAP0	Charge pump capacitor pin 1	N/A	Α
A7	VDBL	Doubled voltage	0	Α
B8	VBAT	Power supply	I	Α
B10	VREG	Regulated supply voltage	0	Α
A9	AGND	Analog ground	N/A	Α
A11	Al4	Audio signal input 4	I	Α
B12	AI2/LOUT2	Audio signal input 2/output signal from preamp 2	I	Α
A13	AI1/LOUT1	Audio signal input 1/output signal from preamp 1	I	Α
B14	AI0/LOUT0	Audio signal input 0/output signal from preamp 0	1	Α
D14	GPIO[4]/LSAD[4]	General-purpose I/O 4/low speed AD input 4	I/O	A/D
E13	GPIO[3]/LSAD[3]	General-purpose I/O 3/low speed AD input 3	I/O	A/D
C13	GPIO[2]/LSAD[2]	General-purpose I/O 2/low speed AD input 2	I/O	A/D
D12	GPIO[1]/LSAD[1]/UART-RX	General-purpose I/O 1/low speed AD input 1/and UART RX	I/O	A/D
E11	GPIO[0]/UART-TX	General-purpose I/O 0/UART TX	I/O	A/D
C9	GNDC	Digital ground	N/A	Α
C11	SDA	I2C data	I/O	D
D10	SCL	I2C clock	I/O	D
E9	EXT_CLK	External clock input/internal clock output	I/O	D
D8	VDDC	Core logic power	0	Α
E7	SPI CLK	Serial peripheral interface clock	0	D
C7	SPI_SERI	Serial peripheral interface input	1	D
D6	SPI CS	Serial peripheral interface chip select	0	D
E5	SPI SERO	Serial peripheral interface output	0	D
D4	PCM FR	PCM interface frame	I/O	D
E3	PCM_SERI	PCM interface input	1	D
D2	PCM_SERO	PCM interface output	0	D
C1	PCM_CLK	PCM interface clock	I/O	D
E1	Reserved	Reserved		

### 2.2 Recommended Design Guidelines

BelaSigna 300 is designed to allow both digital and analog processing in a single system. Due to the mixed-signal nature of this system, the careful design of the printed circuit board (PCB) layout is critical to maintain the high audio fidelity of BelaSigna 300. To avoid coupling noise into the audio signal path, keep the digital traces away from the analog traces. To avoid electrical feedback coupling, isolate the input traces from the output traces.

### 2.2.1. Recommended Ground Design Strategy

The ground plane should be partitioned into two: the analog ground plane (AGND) and the digital ground plane (DGND). These two planes should be connected together at a single point, known as the star point. The star point should be located at the ground terminal of a capacitor on the output of the power regulator as illustrated in Figure 2.



The DGND plane is used as the ground return for digital circuits and should be placed under digital circuits. The AGND plane should be kept as noise-free as possible. It is used as the ground return for analog circuits and it should surround analog components and pins. It should not be connected to or placed under any noisy circuits such as RF chips, switching supplies or digital pads of BelaSigna 300 itself. Analog ground returns associated with the audio output stage should connect back to the star point on separate individual traces.

For details on which signals require special design consideration, see Table 2 and Table 3.

In some designs, space constraints may make separate ground planes impractical. In this case a star configuration strategy should be used. Each analog ground return should connect to the star point with separate traces.

### 2.2.2. Internal Power Supplies

Power management circuitry in BelaSigna 300 generates separate digital (VDDC) and analog (VREG, VDBL) regulated supplies. Each supply requires an external decoupling capacitor, even if the supply is not used externally. Decoupling capacitors should be placed as close as possible to the power pads. Further details on these critical signals are provided in Table 2. Non-critical signals are outlined in Table 3.

Table 2: Critical Signals

Table 2. Official Olgitals		
Pin Name		
VBAT	Power supply	Place 1µF (min) decoupling capacitor close to pin. Connect negative terminal of capacitor to DGND plane.
VREG, VDBL	Internal regulator for analog sections	Place separate 1µF decoupling capacitors close to each pin. Connect negative capacitor terminal to AGND. Keep away from digital traces and output traces. VREG may be used to generate microphone bias. VDBL shall not be used to supply external circuitry.
AGND	Analog ground return	Connect to AGND plane.
VDDC	Internal regulator for digital core	Place 10µF decoupling capacitor close to pin. Connect negative terminal of capacitor to DGND.
GNDC	Digital ground return	Connect to digital ground.
AI0/LOUT0, AI1/LOUT1, AI2/LOUT2, AI4	Audio inputs	Keep as short as possible. Keep away from all digital traces and audio outputs. Avoid routing in parallel with other traces. Connect unused inputs to AGND.
RCVR+, RCVR-, RCVR_HP+, RCVR_HP-	Direct digital audio output	Keep away from analog traces, particularly audio inputs. Corresponding traces should be of approximately the same length. Ideally, route lines parallel to each other.
GNDRCVR	Output stage ground return	Connect to star point. Keep away from all analog audio inputs.
EXT_CLK	External clock input / internal clock output	Minimize trace length. Keep away from analog signals. If possible, surround with digital ground.

Table 3: Non-Critical Signals

Table 3. Non-Childa Signais		
Pin Name	Description	Routing Guideline
CAP0, CAP1	Internal charge pump - capacitor connection	Place 100nF capacitor close to pins
SDA, SCL	I2C port	Keep as short as possible
GPIO[30]	General-purpose I/O	Not critical
UART_RX, UART_TX	General-purpose UART	Not critical
PCM_FRAME, PCM_CLK, PCM_OUT, PCM_IN	Pulse code modulation port	Keep away from analog input lines
LSAD[41]	Low-speed A/D converters	Not critical
SPI_CLK, SPI_CS, SPI_SERI, SPI_SERO	Serial peripheral interface port Connect to EEPROM	Keep away from analog input lines

### 2.2.3. Audio Inputs

The audio input traces should be as short as possible. The input impedance of each audio input pad (e.g., Al0, Al1, Al2, Al4) is high (approximately  $500k\Omega$ ); therefore a 10nF capacitor is sufficient to decouple the DC bias. This capacitor and the internal resistance form a first-order analog high pass filter whose cutoff frequency can be calculated by  $f_{3dB}(Hz) = 1/(R \cdot C \cdot 2\pi)$ , which results with ~30Hz for 10nF capacitor. This 10nF capacitor value applies when the preamplifier is being used, in other words, when a non-unity gain is applied to the signals. When the preamplifier is by-passed, the impedance is reduced; hence, the cut-off frequency of the resulting high-pass filter could be too high. In such a case, the use of a 30-40nF serial capacitor is recommended. In cases where line-level analog inputs without DC bias are used, the capacitor may be omitted for transparent bass response.

BelaSigna 300 provides microphone power supply (VREG) and ground (AGND). Keep audio input traces strictly away from output traces.

Digital outputs (RCVR) MUST be kept away from microphone inputs to avoid cross-coupling.

### 2.2.4. Audio Outputs

The audio output traces should be as short as possible. The trace length of RCVR+ and RCVR- should be approximately the same to provide matched impedances.

### 2.2.5. PCB Manufacturing

For PCB manufacture with BelaSigna 300 WLCSP, ON Semiconductor recommends solder-on-pad (SoP) surface finish. With SoP, the solder mask opening should be non-solder mask-defined (NSMD) and copper pad geometry will be dictated by the PCB vendor's design requirements.

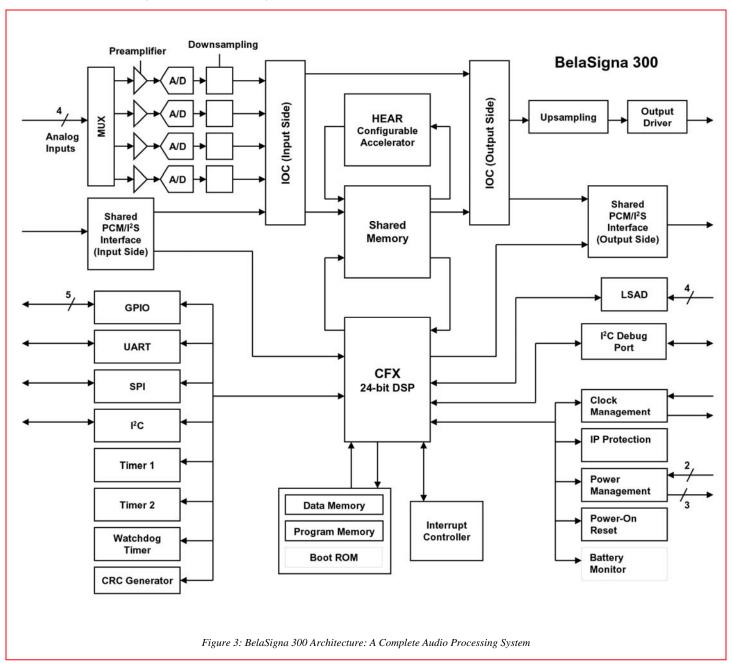
Alternative surface finishes are ENiG and OSP; volume of screened solder paste (#5) should be less than 0.0008mm^3. If no prescreening of solder paste is used, then following conditions must be met:

- (i) the solder mask opening should be >0.3mm in diameter,
- (ii) the copper pad will have 0.25mm diameter, and
- (iii) soldermask thickness should be less than 1mil thick above the copper surface.

ON Semiconductor can provide BelaSigna 300 WLCSP land pattern CAD files to assist your PCB design upon request.

### 3.0 Architecture Overview

The architecture of BelaSigna 300 is shown in Figure 3.



### 3.1 CFX DSP Core

The CFX DSP is a user-programmable general-purpose DSP core that uses a 24-bit fixed-point, dual-MAC, dual-Harvard architecture. It is able to perform two MACs, two memory operations and two pointer updates per cycle, making it well-suited to computationally intensive algorithms.

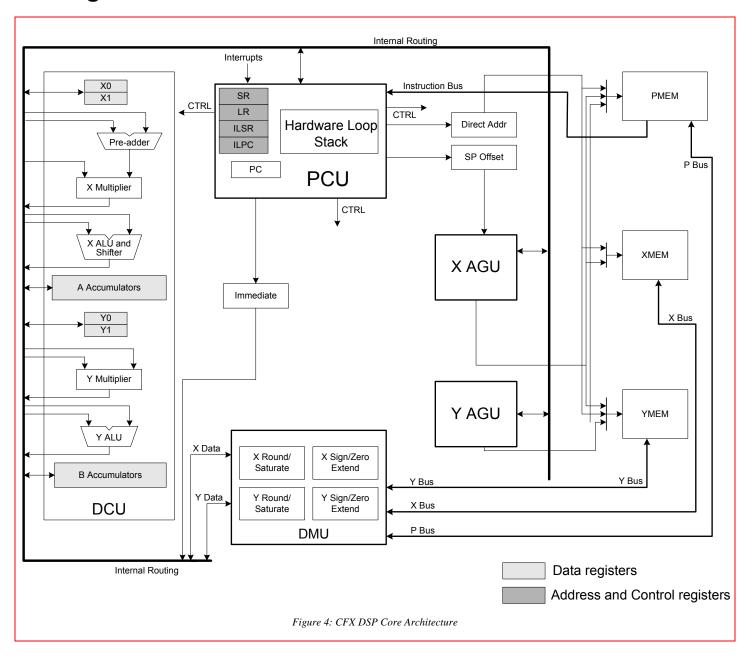
#### The CFX features:

- Dual-MAC 24-bit load-store DSP core
- Four 56-bit accumulators
- Four 24-bit input registers
- Support for hardware loops nested up to 4 deep
- Combined XY memory space (48-bits wide)
- Dual address generator units
- Wide range of addressing modes:
  - Direct
  - o Indirect with post-modification
  - o Modulo addressing
  - Bit reverse

### 3.1.1. CFX DSP Architecture

The CFX architecture encompasses various memory types and sizes, peripherals, interrupt controllers, and interfaces. Figure 4 illustrates the basic architecture of the CFX. The control lines shown exiting the PCU indicate that control signals go from the PCU to essentially all other parts of the CFX.

The CFX employs a parallel instruction set for simultaneous control of multiple computation units. The DSP can execute up to four computation operations in parallel with two data transfers (including rounding and/or saturation as well as complex address updates), while simultaneously changing control flow.



### 3.2 CFX DSP Instruction Set

Table 4 shows the list of all general CFX instructions and their description. Many instructions have multiple variations not shown in the table. Please refer to the CFX DSP Architecture Manual for more details.

Table 4: CFX Summary Instruction Set

DOD Add values (various combinations of accumulators, pointers and data registers)  DOMULADD  Add wo XY data registers, multiply the result by a third XY data register, and store the result in an accumulator Add two XY data registers, multiply the result by a third XY data register, and add the result to an accumulator Add two XY data registers, multiply the result by a third XY data register, negate the result and store it in an accumulator Add two XY data registers or accumulators and shift right one bit, storing the result by a third XY data register, and subtract the result from an accumulator Add two data registers or accumulators and shift right one bit, storing the result and store it in an accumulator Add two data registers or accumulators and shift right one bit, storing the result and accumulator Add two data register or accumulators and shift right one bit, storing the result from an accumulator Add two data register or accumulators and shift right one bit, storing the result from an accumulator Add two data register or accumulators and shift right one bit is storing the result from an accumulator Disposal Storing the register or accumulator and accumul	Table 4: CFX Sun	nmary Instruction Set
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Test a bit in a data register REAKPOINT Halts the DSP for debugging if software breakpoints are enabled through the debug port CALL Call a subroutine  CREAT a word of X memory specified by an X pointer, with update Cher a word of X memory specified by an X pointer, with update Compare a data register or accumulator to another data register or accumulator or a value MPU Compare a data register to a value or another data register or accumulator or a value MPU Compare a data register to a value or another data register as unsigned values or compare two accumulators as unsigned values DIVIST Division step for dividing data register and stores the result to a data register NDLOOP End a hardware loop before the count has reached zero Software interrupt Software interrupt CAD CAGLABS Calculate the logarithm base 2 of the absolute value of a data register, accumulator or data Calculate the logarithm base 2 of the absolute value of a data register, storing the result in a data register or accumulator AIX Determine the maximum value of two data registers or accumulators and store the result in a data register or accumulator AIX Determine the minimum value of two data registers or accumulators and store the result in a data register or accumulator AIVE Multiply two XY data registers, storing the result in an accumulator MULNEG Multiply two XY data registers, and add the result to an accumulator MULNEG Multiply two XY data registers, negate the result from an accumulator MULNEG Multiply two XY data registers, negate the result and store the result, and store the result in a data register No peration  RETURN Return from a bitwise OR operation on two accumulators storing the result in a data register or accumulator on two data registers or a data register RETURN Return from a subroutine RETURN Return from an interrupt SHRLL Shift a data register right arithmetically SHRL Shift a data register right arithmetically SHRL Shift a data register right arithmetically SHRL Shift a data register right arithmetically SUBMULNEG Subract	BITSET	Set a bit in the register
REAKPOINT  ALL  Call a subroutine  Clear a word of X memory specified by an X pointer, with update  Clear a word of X memory specified by an X pointer, with update  Clear a word of X memory specified by an X pointer, with update  MP  Compare a data register to a value or another data register or accumulator or a value  Compare a data register to a value or another data register as unsigned values or compare two accumulators as unsigned values  INST  Division step for dividing data register by data register and stores the result to a data register  End a hardware loop before the count has reached zero  DIVIST  Software interrupt  OAD  Load a register, accumulator or a memory location with another register, accumulator or data  COG2ABS  Calculate the logarithm base 2 of the absolute value of a data register, storing the result in a data register or accumulator  AXX  Determine the maximum value of two data registers or accumulators and store the result in a data register or accumulator  AXX  Determine the minimum value of two data registers or accumulators and store the result in a data register or accumulator  MIN  Determine the minimum value of two data registers or accumulators and store the result in a data register or accumulator  MULNEG  Multiply two XY data registers, storing the result in an accumulator  MULNEG  Multiply two XY data registers, and add the result to an accumulator  MULNEG  Multiply two XY data registers, and subtract the result from an accumulator  MULSUB  Multiply two XY data registers, and subtract the result from an accumulator  No operation  No operation  Perform a bitwise OR operation on two accumulators storing the result in an active result in a data register or accumulator in a data register or accumulator storing the result in an accumulator or on two data registers or a data register or accumulator in a data register or accumulator or on two data registers or a data register or accumulator storing the result in a data register or accumulator or on two data registers or accumulator	BITTGL	Toggle a bit in a data register
Call a subroutine Cite of word of X memory specified by an X pointer, with update CMP Compare a data register or accumulator to another data register or accumulator or a value CMPU Compare a data register to a value or another data register or accumulator or a value CMPU Compare a data register to a value or another data register as unsigned values or compare two accumulators as unsigned values DMPU Compare a data register to a value or another data register as unsigned values or compare two accumulators as unsigned values DMPU Compare a data register to a data register or data register or data register CMD.COP End a hardware loop before the count has reached zero DOTO Branch to an address or label NTERRUPT COAD Load a register, accumulator or a memory location with another register, accumulator or data CoC2ABS Calculate the logarithm base 2 of the absolute value of a data register, storing the result in a data register COCP Loop with a specified count AMX Determine the maximum value of two data registers or accumulators and store the result in a data register or accumulator AMX Determine the minimum value of two data registers or accumulators and store the result in a data register or accumulator AMV Move a register or accumulator to a register or accumulator AMUL Multiply two XY data registers, storing the result in an accumulator AMULADD Multiply two XY data registers, and add the result to an accumulator AMULADD Multiply two XY data registers, negate the result from an accumulator AMULADD Multiply two XY data registers, and subtract the result from an accumulator AMULADD Multiply two XY data registers, so accumulator storing the result in a data register or accumulator AMULADD Multiply two XY data registers, so accumulator storing the result in a data register or accumulator AMULADD Multiply two XY data registers, so accumulator storing the result in an accumulator AMULADD Multiply two XY data registers or accumulators and so accumulator storing the result in an accumulator or not two data registers or accu	BITTST	Test a bit in a data register
Clear a word of X memory specified by an X pointer, with update CMP Compare a data register or accumulator to another data register or accumulator or a value Compare a data register to a value or another data register as unsigned values or compare two accumulators as unsigned values Division step for dividing data register by data register and stores the result to a data register End a hardware loop before the count has reached zero DOTO Branch to an address or label NTERRUPT Software interrupt Load a register, accumulator or a memory location with another register, accumulator or data Calculate the logarithm base 2 of the absolute value of a data register, storing the result in a data register Loop with a specified count AXX Determine the maximum value of two data registers or accumulators and store the result in a data register or accumulator MIN Determine the minimum value of two data registers or accumulators and store the result in a data register or accumulator MIVE Move a register or accumulator to a register or accumulator MULADD Multiply two XY data registers, storing the result in an accumulator MULADD Multiply two XY data registers, and soft the result to an accumulator MULSUB Multiply two XY data registers, and subtract the result from an accumulator MULSUB Nultiply two XY data registers, and subtract the result from an accumulator MOP No operation Perform a bitwise OR operation on two accumulators storing the result in an accumulator or no two data registers or a data register RETURN Return from a subroutine RETURN Return from	BREAKPOINT	Halts the DSP for debugging if software breakpoints are enabled through the debug port
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Compare a data register or accumulator to another data register or accumulator or a value Compare a data register to a value or another data register as unsigned values or compare two accumulators as unsigned values Division step for dividing data register by data register and stores the result to a data register End a hardware loop before the count has reached zero Barnch to an address or label NTERRUPT Software interrupt COAD Load a register, accumulator or a memory location with another register, accumulator or data COG2ABS Calculate the logarithm base 2 of the absolute value of a data register, accumulator or data COG2ABS Calculate the logarithm base 2 of the absolute value of a data register, storing the result in a data register COP Loop with a specified count DAX Determine the maximum value of two data registers or accumulators and store the result in a data register or accumulator MIN Determine the minimum value of two data registers or accumulators and store the result in a data register or accumulator MIVE Move a register or accumulator to a register or accumulator MIVE MULNEG Multiply two XY data registers, storing the result in an accumulator MIVE MULNEG Multiply two XY data registers, and add the result on an accumulator MIVE MULNEG Multiply two XY data registers, and subtract the result from an accumulator MIVE MULOG2ABS Calculate the logarithm base 2 of the absolute value of a data register or accumulator No operation Perform a bitwise OR operation on two accumulators storing the result in an accumulator or on two data registers or a data register METURN Return from a subroutine RETURN Return from an interrupt SHILL Shift a data register right arithmetically SHILL Shift a data register inght arithmetically SHILL Shift a data register inght raithmetically SHILL Shift a data register inght price in interrupt and then wake up from sleep mode SUBMULADD Subtract two XY data registers, multiply the result by a third XY data register, and store the result in an accumulator SUBMULSUB SUBMULADD Subtract t	CLR	Clear a word of X memory specified by an X pointer, with update
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	XOR	Perform a bitwise XOR operation on two data registers or a data register and a value, storing the result in a data register

### 3.3 HEAR Configurable Accelerator

The HEAR Configurable Accelerator is a highly optimized signal processing engine that is configured through the CFX. It offers high speed, high flexibility and high performance, while maintaining low power consumption. For added computing precision, the HEAR supports block floating point processing. Configuration of the HEAR is performed using the HEAR configuration tool (HCT). For further information on the usage of the HEAR and the HCT, please refer to the HEAR Configurable Accelerator Reference Manual.

The HEAR is optimized for advanced audio algorithms, including but not limited to the following:

- Dynamic range compression
- Directional processing
- Acoustic echo cancellation
- Noise reduction

To provide the ability for these algorithms to be executed efficiently, the HEAR excels at the following:

- Processing using a weighted overlap add (WOLA) filterbank or FFT
- Time domain filtering
- Subband filtering
- Attack/release filtering
- Vector addition/subtraction/multiplication
- Signal statistics (such as average, variance and correlation)

### 3.4 Input/Output Controller (IOC)

The IOC is responsible for the automated data moves of all audio samples transferred in the system. The IOC can manage any system configuration and route the data accordingly. It is an advanced audio DMA unit.

### 3.5 Memory

### 3.5.1. RAM & ROM

The size and width of each of the RAM and ROM structures are shown in Table 5:

Table 5:RAM and ROM Structure

Table on a milana real and a constant						
Memory Structure	Data Width	Memory Size				
Program memory (ROM)	32	2048				
Program memory (RAM)	32	12288				
X memory (RAM)	24	6144				
Math library LUT (ROM)	24	128				
Y memory (RAM)	24	2048				

### 3.5.2. Shared Memories

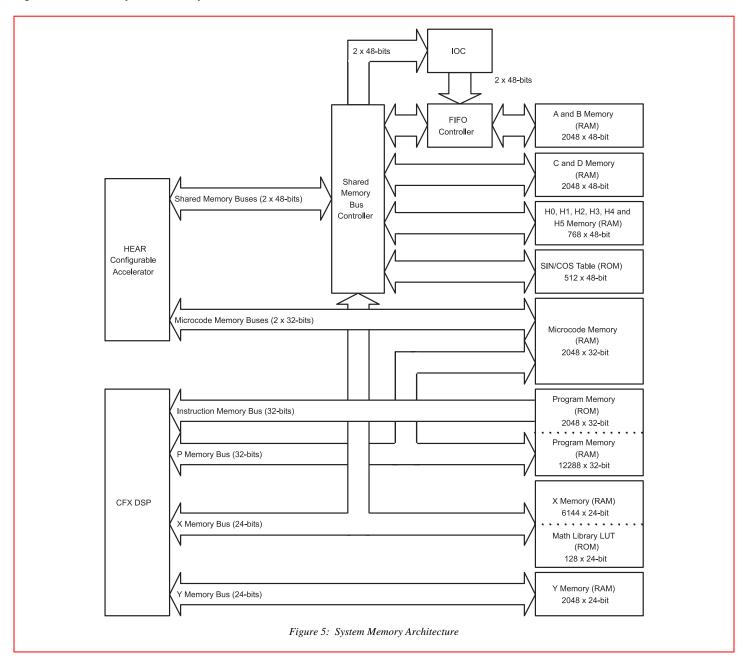
The shared CFX/HEAR memories include the following:

Table 6: Shared Memories

Туре	Name	Size
Data memory (RAM)	H0MEM, H1MEM, H2MEM, H3MEM, H4MEM, H5MEM	Each 128x48-bit words
FIFO memory (RAM)	AMEM, BMEM	Each 1024x48-bit words
Coefficient memory (RAM)	CMEM, DMEM	Each 1024x48-bit words
Data ROM	SIN/COS LUT	512x48-bit words containing the 512 point sin/cos look up table
Microcode memory (RAM)	MICROCODE_MEM	2048x32-bit words

### 3.5.3. Memories Structure

Figure 5 shows the system memory structure. The individual blocks are described in the sections that follow.

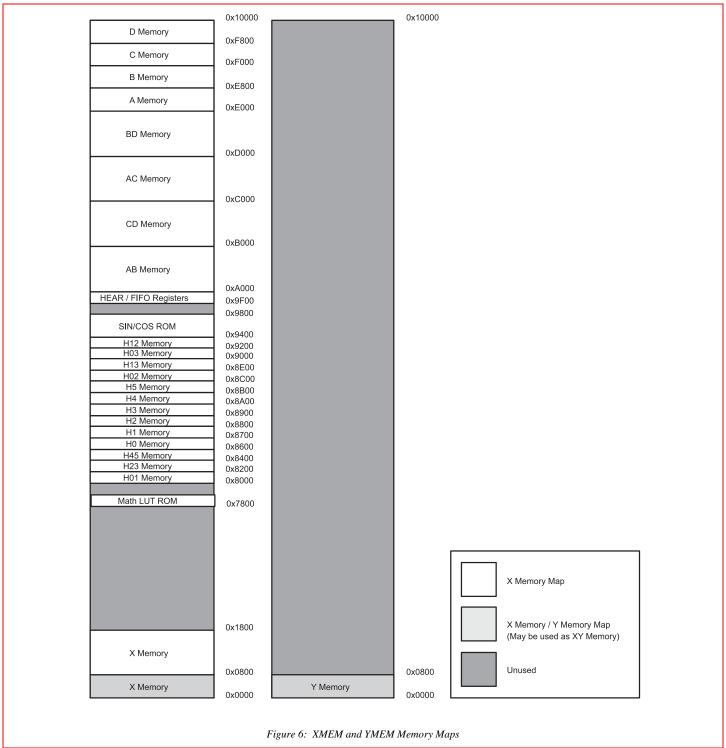


### 3.5.4. FIFO Controller

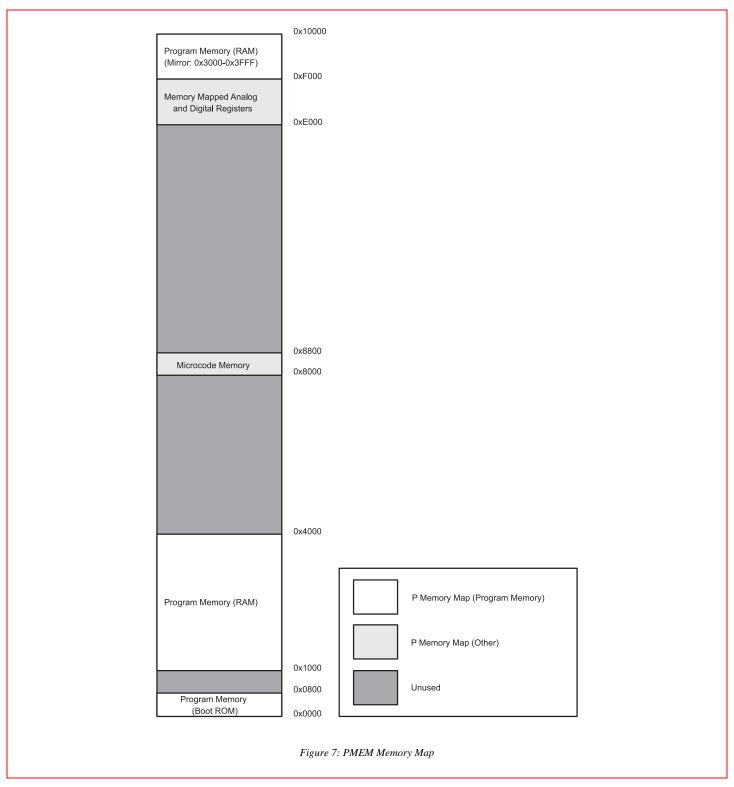
The FIFO controller handles the moving of data to and from the FIFOs, after being initially configured. Up to eight FIFOs can be created by the FIFO controller, four in A memory (AMEM) and four in B memory (BMEM). Each FIFO has a block counter that counts the number of samples read or written by the IOC. It creates a dedicated interrupt signal, updates the block counter and updates the FIFO pointers when a new block has been read or written.

### 3.5.5. Memory Maps

The structure of the XMEM and YMEM address spaces are shown in Figure 6.



The structure of the PMEM address space is shown in Figure 7.



### 3.6 Other Digital Blocks and Functions

### 3.6.1. General-Purpose Timer

The CFX DSP system contains two general-purpose timers. These can be used for scheduling tasks that are not part of the sample-based signal-processing scheme, such as checking the battery voltage, and periodically asserting the available analog and digital inputs for purposes such as reading the value of a volume control potentiometer or detecting input from a push button.

### 3.6.2. Watchdog Timer

The watchdog timer is a programmable hardware timer that operates from the system clock and is used to ensure system sanity. It is always active and must be periodically acknowledged as a check that an application is still running. Once the watchdog times out, it generates an interrupt. If left to time out a second consecutive time without acknowledgement, a system reset will occur.

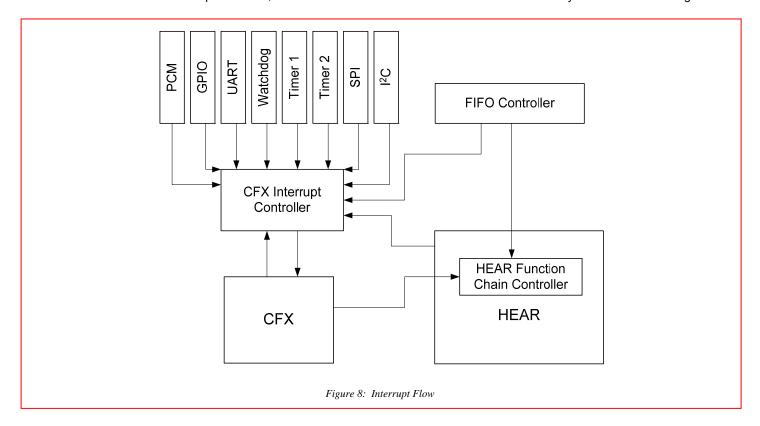
### 3.6.3. Interrupts

The interrupt flow of the system handles interrupts generated by the CFX DSP core and the HEAR accelerator. The CFX interrupt controller receives interrupts from the various blocks within the system. The FIFO controller can send interrupts to the CFX. The HEAR can generate events which are interrupts in the CFX.

### 3.6.4. Hear Function Chain Controller

The HEAR function chain controller responds to commands from the CFX, and events from the FIFO controller. It must be configured by the CFX to enable the triggering of particular function chains within a microcode configuration. This is accomplished through the appropriate setting of control registers as described in the Hardware Reference Manual for BelaSigna 300.

The interaction between the interrupt controller, the HEAR function chain controller and the rest of the system are shown in Figure 8.



### 3.6.5. Algorithm and Data Security

Algorithm software code and user data that requires permanent retention is stored off the BelaSigna 300 chip in separate non-volatile memory. To support this, the BelaSigna 300 chip can gluelessly interface to an external SPI EEPROM.

To prevent unauthorized access to the sensitive intellectual property (IP) stored in the EEPROM, a comprehensive system is in place to protect manufacturer's application code and data. When locked the system implements an access restriction layer that prevents access to both volatile and non-volatile system memory. When unlocked, both memory and EEPROM are accessible.

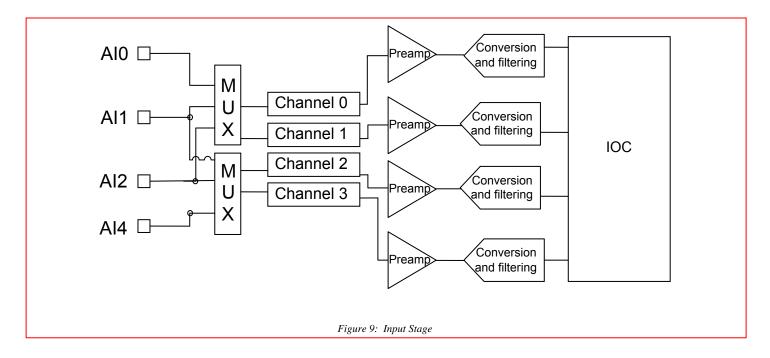
To protect the IP in the non-volatile memory the system supports decoding algorithm and data sections belonging to an application that have been encrypted using the advanced encryption standard (AES) and stored in non-volatile memory. While system access restrictions are in place, the keys used in the decryption of these sections will be secured from external access by the regular access restrictions. When the system is externally "unlocked" these keys will be cleared, preventing their use in decoding an application by non-authorized parties. After un-restricting access in this way the system may then be restored by re-programming the decryption keys.

### 3.7 Analog Blocks

### 3.7.1. Input Stage

The analog audio input stage is comprised of four individual channels. For each channel, one input can be selected from any of the four possible input sources and is then routed to the input of the programmable preamplifier that can be configured for bypass or gain values of 12 to 30dB (3dB steps). The input stage is shown in Figure 9.

A built-in feature allows a sampling delay to be configured for any one or more channels. This is useful in beam-forming applications.



### 3.7.2. Input Dynamic Range Extension (IDRX)

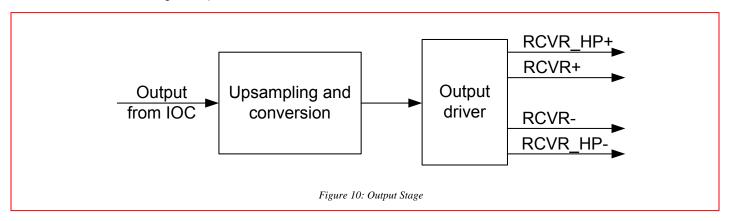
To increase the input dynamic range for a particular application, it is possible to pair-wise combine the four AD converters found on BelaSigna 300. This will increase the dynamic range up to 110dB. When this technique is used, the device handles the preamplifier gain configuration based on the input level and sets it in such a way as to give the maximum possible dynamic range. This avoids having to make the design trade-off between sufficient amplification for low-level signals and avoiding saturation for high-level signals.

### 3.7.3. Output Stage

The first part of the output stage interpolates the signal for highly oversampled D/A conversion and automatically selects the oversampling rate. The signal is then routed to the differential direct-digital output driver.

The low-impedance direct-digital output is driven by a pulse-density modulated output and can be used to directly drive an output transducer without the need for a separate power amplifier. The output stage is shown in Figure 10.

BelaSigna 300 has an option for high-power mode that decreases the impedance of the output stage, thus permitting higher possible acoustic output levels. To use this feature, RCVR\_HP+ and RCVR+ should be tied together before connecting to the output transducer, and RCVR\_HP- and RCVR- should be tied together. The appropriate registers must also be configured (as specified in the Hardware Reference Manual for BelaSigna 300).



### 3.7.4. Clock Generation Circuitry

BelaSigna 300 is equipped with an un-calibrated internal RC oscillator that will provide clock support for booting and stand-by mode operations. This internal clocking circuitry cannot be used during normal operation. An external clock signal must be present on the EXT\_CLK pin to allow BelaSigna 300 to operate. All other needed clocks in the system are derived from this external clock frequency.

### 3.8 Power Supply Unit

### 3.8.1. Voltage Mode

BelaSigna 300 operates in: **Low voltage (LV) power supply mode**. This mode allows integration into a wide variety of devices with a range of voltage supplies and communications levels. BelaSigna 300 operates from a nominal supply of 1.8V on VBAT, but this can scale depending on available supply. The digital logic runs on an internally generated regulated voltage, in the range of 0.9V to 1.2V. All digital I/O pads including the SPI port run from the same voltage as supplied on VBAT.

The power management on BelaSigna 300 includes the power-on-reset (POR) functionality as well as power supervisory circuitry. These two components work together to ensure proper device operation under all battery conditions.

The power supervisory circuitry monitors both the battery supply voltage (VBAT) and the internal digital supply voltage (VDDC). This circuit is used to start the system when VBAT reaches a safe startup voltage, and to reset the system when either of the VBAT or VDDC voltages drops below a relevant voltage threshold. The relevant threshold voltages are shown in Table 7.

Table 7: Power Management Thresholds

Threshold	Voltage Level
VBAT monitor startup	0.70V
VBAT startup	0.82V +/- 50mV
VBAT and VDDC shutdown	0.80V +/- 50mV

### 3.8.2. Power-on-Reset (POR) and Booting Sequence

BelaSigna 300 uses a POR sequence to ensure proper system behavior during start-up and proper system configuration after start-up. At the start of the POR sequence, the audio output is disabled and all configuration and control registers are asynchronously reset to their default values (as specified in the Hardware Reference Manual for BelaSigna 300). All CFX DSP registers are cleared and the contents of all RAM instances are unspecified at this point.

The POR sequence consists of two phases: voltage supply stabilization and boot ROM initialization. During the voltage supply stabilization phase, the following steps are performed:

- 1. The internal regulators are enabled and allowed to stabilize.
- 2. The internal charge pump is enabled and allowed to stabilize.
- 3. SYSCLK is connected to all of the system components.
- 4. The system switches to external clocking mode

### 3.8.3. Power Management Strategy

BelaSigna 300 has a built-in power management unit that guarantees valid system operation under any voltage supply condition to prevent any unexpected audio output as the result of any supply irregularity. The unit constantly monitors the power supply and shuts down all functional units (including all units in the audio path) when the power supply voltage goes below a level at which point valid operation can no longer be guaranteed.

Once the supply voltage rises above the startup voltage of the internal regulator that supplies the digital subsystems (VDDC<sub>STARTUP</sub>) and remains there for the length of time  $T_{POR}$ , a POR will occur. If the supply is consistent, the internal system voltage will then remain at a fixed nominal voltage (VDDC<sub>NOMINAL</sub>). If a spike occurs that causes the voltage to drop below the shutdown internal system voltage (VDDC<sub>SHUTDOWN</sub>), the system will shut down. If the voltage rises again above the startup voltage and remains there for the length of time  $T_{POR}$ , a POR will occur. If operating directly off a battery, the system will not power down until the voltage drops below the VDDC<sub>SHUTDOWN</sub> voltage as the battery dies. This prevents unwanted resets when the voltage is just on the edge of being too low for the system to operate properly because the difference between VDDC<sub>STARTUP</sub> and VDDC<sub>SHUTDOWN</sub> prevents oscillation around the VDDC<sub>SHUTDOWN</sub> point.

#### 3.9 Other Analog Support Blocks and Functions

### 3.9.1. Low-Speed A/D Converters (LSAD)

The BelaSigna 300 chip has four LSAD channels that connect to external analog inputs for purposes such as for reading the value of a potentiometer or an analog sensor (LSAD[1..4]). The native data format for the LSAD is 10-bit two's-complement. However, a total of eight operation modes are provided that allow a configurable input dynamic range in cases where certain minimum and maximum values for the converted inputs are desired, such as in the case of a volume control where only input values up to a certain magnitude are allowed. Each LSAD channel is sampled at a nominal frequency of 1.6kHz when using the default settings. Each LSAD pin is multiplexed with a GPIO function (see 3.10.1) as such the functionality of the pin can be either a GPIO or an LSAD depending on the configuration.

#### 3.9.2. Battery Monitor

A programmable on-chip battery monitor is available for overall system power management. The battery monitor works by incrementing a counter value every time the battery voltage goes below a desired, configurable threshold value. This counter value can be used in an application-specific power-management algorithm running on the CFX. The CFX can initiate any desired actions once the battery hits a predetermined value.

### 3.10 Digital Interfaces

### 3.10.1. General-Purpose Input Output (GPIO) Ports

BelaSigna 300 has five GPIO ports that can connect to external digital inputs such as push buttons, or digital outputs such as the control or trigger of an external companion chip (GPIO[0..4]). The direction of these ports (input or output) is configurable and each pin has an internal pull-up resistor when configured as a GPIO. A read from an unconnected pin will give a value of logic 1. Four of the five GPIO pins are multiplexed with an LSAD (see 3.9.1) and as such the functionality of the pin can be either a GPIO or an LSAD depending on the configuration. Note that GPIO0 cannot be used as an LSAD.

### 3.10.2. Inter-IC Communication (I<sup>2</sup>C) Interfaces

The I<sup>2</sup>C interface is an industry-standard interface that can be used for high-speed transmission of data between BelaSigna 300 and an external device. The interface operates at speeds up to 400Kbit/sec for system clocks (EXT\_CLK) higher than 1.28MHz. In product development mode, the I<sup>2</sup>C interface is used for application debugging purposes, communicating with the BelaSigna 300 development tools. The interface can be configured to operate in either master mode or slave mode.

### 3.10.3. Serial Peripheral Interface (SPI) Port

An SPI port is available on BelaSigna 300 for applications such as communication with a non-volatile memory (EEPROM). The I/O levels on this port are defined by the voltage on the VBAT pin. The SPI port operates in master mode only, which supports communications with slave SPI devices.

### 3.10.4. PCM Interface

BelaSigna 300 includes a highly configurable pulse code modulation (PCM) interface that can be used to stream signal, control and configuration data into and out of the device. The I/O levels on this port are defined by the voltage on the VBAT pin.

### 3.10.5. UART Interface

A general-purpose two-pin UART interface is available for RS-232 compatible communications. The baud rate (bits/second) of this interface is typically configurable within a range of 0.4 to 320 kbps, depending on the application's system clock. The I/O levels on this port are defined by the voltage on the VBAT pin.

### 4.0 Figures and Data

### 4.1 Absolute Maximum Ratings

Table 8: Absolute Maximum Ratings

Parameter	Min.	Max.	Unit				
Voltage at any input pin	-0.3	2.0	V				
Operating supply voltage	0.9	2.0	V				
Operating temperature range	-40	85	°C				
Storage temperature range	-40	85	°C				
Caution: Class 2 ESD Sensitivity, JESD22-A114-B (2000V)							

### **4.2 Electrical Performance Specifications**

The tests were performed at 20°C with a clean 1.8V supply voltage. BelaSigna 300 was running in low voltage mode (VDDC=1.3V). The system clock (SYS\_CLK) was set to 5.12MHz and the sampling frequency is 16kHz unless otherwise noted.

Parameters marked as screened are tested on each chip. Other parameters are qualified but not tested on every part.

Table 9: Electrical Specifications

Overall								
Description	Symbol	Conditions	Min.	Тур.	Max.	Units	Screened	
Supply voltage	$V_{BAT}$		0.9	1.8	2.0	V		
Current consumption	I <sub>BAT</sub>	Filterbank, 100% CFX usage, 5.12MHz, 16kHz	-		-	μA	<b>V</b>	
		WDRC, 1.8V Excludes output drive current	-	600	-	μA		
		AEC, 1.8V Excludes output drive current	-	2.1	-	mA		
		Theoretical maximum Excludes output drive current	-	10	-	mA		
		Deep Sleep current Ambient room temperature	-	26	40	μA		

VREG (1μF External Capacitor)							
Description	Symbol	Conditions	Min.	Тур.	Max.	Units	Screened
Regulated voltage output	V <sub>REG</sub>		0.95	1.00	1.05	V	V
Regulator PSRR	V <sub>REG_PSRR</sub>	1kHz	50	55	-	dB	
Load current	I <sub>LOAD</sub>		-	-	2	mA	
Load regulation	LOAD <sub>REG</sub>		-	6.1	6.5	mV/mA	
Line regulation	LINE <sub>REG</sub>		-	2	5	mV/V	

Table 10: Electrical Specification (Continued)

VDBL (1μF External Capacitor)							
Description	Symbol	Conditions	Min.	Тур.	Max.	Units	Screened
Regulated doubled voltage output	VDBL		1.9	2.0	2.1	V	V
Regulator PSRR	VDBL <sub>PSRR</sub>	1kHz	35	41	-	dB	
Load current	I <sub>LOAD</sub>		-	-	2.5	mA	
Load regulation	LOAD <sub>REG</sub>		-	7	10	mV/mA	√
Line regulation	LINE <sub>REG</sub>		-	10	20	mV/V	

VDDC (1µF External Capacitor)							
Description	Symbol	Conditions	Min.	Тур.	Max.	Units	Screened
Digital supply voltage output	VDDC	Configured by a control register	0.79	0.95	1.2	V	V
VDDC output level adjustment	VDDC <sub>STEP</sub>		27	29	31	mV	
Regulator PSRR	VDDC <sub>PSRR</sub>	1kHz	25	25.5	26	dB	
Load current	I <sub>LOAD</sub>		-	-	3.5	mA	
Load regulation	LOAD <sub>REG</sub>		-	3	12	mV/mA	<b>V</b>
Line regulation	LINE <sub>REG</sub>		-	3	8	mV/V	

Power-on-Reset (POR)							
Description	Symbol	Conditions	Min.	Тур.	Max.	Units	Screened
POR startup voltage	VDDC <sub>STARTUP</sub>		0.775	0.803	0.837	V	
POR shutdown voltage	VDDC <sub>SHUTDOWN</sub>		0.755	0.784	0.821	V	
POR hysteresis	POR <sub>HYSTERESIS</sub>		13.8	19.1	22.0	mV	
POR duration	T <sub>POR</sub>		11.0	11.6	12.3	ms	

Table 10: Electrical Specification (Continued)

Input Stage							
Description	Symbol	Conditions	Min.	Тур.	Max.	Units	Screened
Analog input voltage	V <sub>IN</sub>		0	-	2	V	
Preamplifier gain tolerance	PAG	1kHz	-1	0	1	dB	$\checkmark$
Innut impodonos	В	0dB preamplifer gain	-	239	-	kΩ	
Input impedance	R <sub>IN</sub>	Non-zero preamplifier gains	550	578	615	kΩ	√
Input referred noise	IN <sub>IRN</sub>	Unweighted, 100Hz to 10kHz BW Preamplifier setting: 0dB 12dB 15dB 18dB 21dB 24dB 27dB 30dB	- - - - - - -	39 10 7 6 4.5 4 3.5 3	50 12 9 8 5.5 5 4.5	μVrms	
Input dynamic range	IN <sub>DR</sub>	1 kHz, 20Hz to 8kHz BW Preamplifier setting: 0dB 12dB 15dB 18dB 21dB 24dB 27dB 30dB	85 84 84 83 82 81 80 78	89 88 88 87 86 85 83 81	- - - - - -	dB	
nput peak THD+N	IN <sub>THDN</sub>	Any valid preamplifier gain, 1kHz	-	-70	-63	dB	<b>V</b>

Direct Digital Output							
Description	Symbol	Conditions	Min.	Тур.	Max.	Units	Screened
Maximum load current	I <sub>DO</sub>	Normal mode	-	-	50	mA	
Output impedance	R <sub>DO</sub>	Normal mode	-	-	5.5	Ω	
Output dynamic range	DO <sub>DR</sub>	Unweighted, 100Hz to 8kHz BW, mono	92	95	-	dB	
Output THD+N	DO <sub>THDN</sub>	Unweighted, 100Hz to 22kHz BW, mono	-	-79	-76	dB	
Output voltage	DO <sub>VOUT</sub>		-V <sub>BATRCVR</sub>		V <sub>BATRCVR</sub>	V	

Anti-Aliasing Filters (Input and Output)							
Description	Symbol	Conditions	Min.	Тур.	Max.	Units	Screened
Preamplifier filter cut-off frequency		Preamp not bypassed	-	20	-	kHz	
Digital anti-aliasing filter cut-off frequency			-	f <sub>s</sub> /2	-		
Passband flatness			-1	-	1	dB	
Input stopband attenuation		60kHz (12kHz cut-off)	-	60	-	dB	

Table 10: Electrical Specification (Continued)

Low-Speed A/D							
Description	Symbol	Conditions	Min.	Тур.	Max.	Units	Screened
Input voltage		Peak input voltage	0	-	2.0	V	V
INL		From GND to 2*VREG	-	4	10	LSB	
DNL		From GND to 2*VREG	-	-	2	LSB	
Maximum variation over temperature (0°C to 50°C)			-	-	5	LSB	
Sampling frequency		All channels sequentially	-	12.8	-	kHz	
Channel sampling frequency		8 channels	-	1.6	-	kHz	

Digital Pads							
Description	Symbol	Conditions	Min.	Тур.	Max.	Units	Screened
Voltage level for high input	V <sub>IH</sub>		VBAT * 0.8	-	-	V	√
Voltage level for low input	V <sub>IL</sub>		-	-	VBAT * 0.2	V	$\checkmark$
Voltage level for high output	V <sub>OH</sub>	2mA source current	VBAT * 0.8	VBAT * 0.87	-	V	√
Voltage level for low output	V <sub>OL</sub>	2mA sink current	-	VBAT * 0.08	VBAT * 0.2	V	$\checkmark$
Input capacitance for digital pads	C <sub>IN</sub>		-	4	-	pF	
Pull-up resistance for digital input pads	R <sub>UP_IN</sub>		-	270	-	kΩ	√
Pull-down resistance for digital input pads	R <sub>DOWN_IN</sub>		-	275	-	kΩ	√
Sample rate tolerance	FS	Sample rate of 16kHz or 32kHz	-1	+/-0	+1	%	
Rise and fall time	Tr, Tf	Digital output pad				ns	
ESD		Human Body Model			2	kV	
		Machine Model			200	V	
Latch-up		V <gndc, v="">VBAT</gndc,>			200	mA	

Oscillation Circuitry							
Description	Symbol	Conditions	Min.	Тур.	Max.	Units	Screened
Internal oscillator frequency	SYS_CLK		0.5	-	10.24	MHz	
Calibrated internal clock frequency	SYS_CLK		-1	+/-0	+1	%	<b>√</b>
Internal oscillator jitter		System clock: 1.28MHz	-	0.4	1	ns	
External assillator talaranasa	EVT CLK	Duty cycle	45	50	55	%	
External oscillator tolerances	EXT_CLK	System clock: 30MHz	-	-	300	ps	
Maximum working frequency	CLK <sub>MAX</sub>	External clock; VBAT: 1.8V		-	40	MHz	

Digital Interfaces							
Description	Symbol	Conditions	Min.	Тур.	Max.	Units	Screened
I2C baud rate		System clock ≥ 1.28MHz	-	-	400	kbps	
General-purpose UART baud rate		System clock ≥ 5.12MHz	-	1	-	Mbps	

### **4.3 Environmental Characteristics**

Parts supplied against this specification will have been qualified as follows:

Table 10: Environmental Characteristics

Table 101 Ellinolitational origination						
Characteristics						
Packaging Level						
Moisture sensitivity level	JEDEC Level 3					
	30°C / 60% RH for 192 hours					
Pressure cooker test (PCT)	121°C / 100% RH / 2 atm for 168 hours					
Thermal cycling test (TCT)	-65°C to 150°C for 1000 cycles					
Highly accelerated stress test (HAST)	130°C / 85% RH for 100 hours					
High temperature stress test (HTST)	150°C for 1000 hours					

Board Level	
Temperature	-40°C to 125°C for 2500 cycles with no failures
Drop	1m height with no failures
Bending	1mm deflection / 2Hz

### **5.0** Assembly Information

### **5.1 Carrier Details**

ON Semiconductor offers tape and reel packing for BelaSigna 300 WLCSP. The packing consists of a pocketed carrier tape, a cover tape, and a molded anti-static polystyrene reel. The carrier and cover tape create an ESD safe environment, protecting the components from physical and electrostatic damage during shipping and handling.

TBD

Figure 11: Package Orientation on Tape

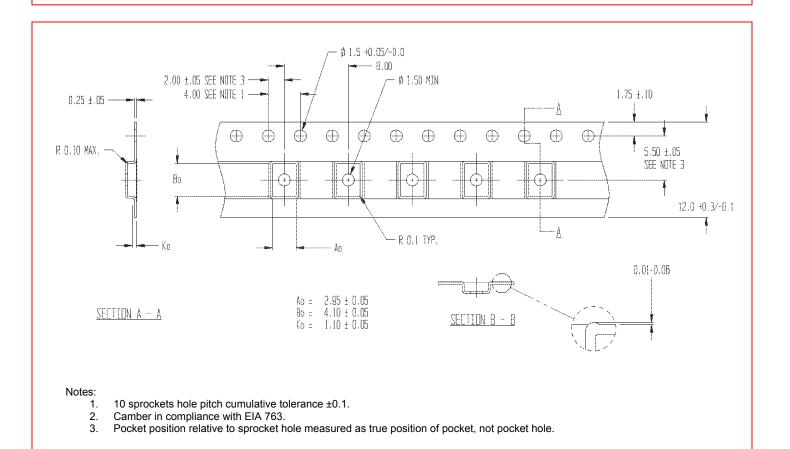


Figure 12: CABGA Carrier Tape Drawing

### 5.1.1. Sample Label Standard

Sample Label will be available Q1 2009.

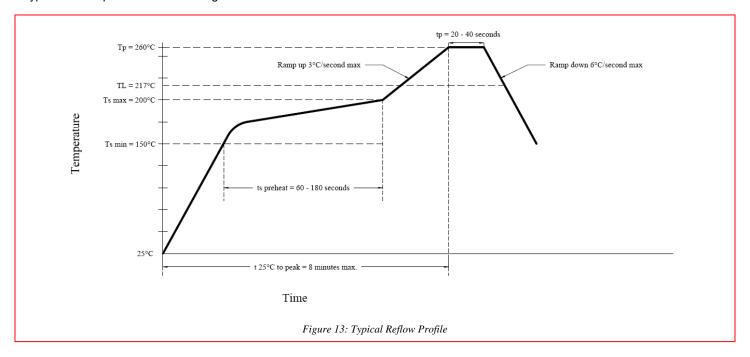
### **5.2 Re-Flow Information**

The re-flow profile depends on the equipment that is used for the re-flow and the assembly that is being re-flowed. Use the following table from the JEDEC Standard 22-A113D and J-STD-020C as a guideline:

Table 11: Re-flow Information

Profile Feature	Pb-free Assembly
Average Ramp-Up Rate (TL to TP)	3°C/second maximum
Preheat	
Temperature minimum (TSMIN)	150°C
Temperature maximum (TSMAX)	200°C
Time (min. to max.) (ts)	60-180 seconds
TSMAX to TL	
Ramp-up rate	3°C/second maximum
Time Maintained Above	
Temperature (TL)	217°C
Time (tL)	60-150 seconds
Peak Temperature (TP)	260 +0/-5°C
Time within 5°C of Actual Peak Temperature	20-40 seconds
Ramp-Down Rate	6°C/second maximum
Time 25°C to Peak Temperature	8 minutes maximum

A typical re-flow profile is shown in Figure 13.



### 5.3 Electrostatic Discharge (ESD) Sensitive Device

**CAUTION:** ESD sensitive device. Permanent damage may occur on devices subjected to high-energy electrostatic discharges. Proper ESD precautions in handling, packaging and testing are recommended to avoid performance degradation or loss of functionality. Device is 2kV HBM ESD qualified.



### **5.4 Moisture Sensitivity Level**

This device will be qualified MSL 3 or better.

### 6.0 Miscellaneous

### **6.1 Ordering Information**

To order BelaSigna 300 WLCSP, please contact your account manager and ask for part number B300W35A102XYG.

### 6.2 Chip Identification

Chip identification information can be retrieved by using the Communications Accelerator Adaptor (CAA) tool along with the protocol software provided by ON Semiconductor (see CAA instruction manual). For BelaSigna 300, the key identifier components and values are as follows:

Chip Family: 0x03Chip Version: 0x02Chip Revision: 0x0100

### 6.3 Support Software

A full suite of comprehensive tools is available to assist software developers from the initial concept and technology assessment through to prototyping and product launch. Simulation, application development and communication tools as well as an Evaluation and Development Kit (EDK) facilitate the development of advanced algorithms on BelaSigna 300.

### 6.4 Training

To facilitate development on the BelaSigna 300 platform, training is available upon request. Contact your account manager for more information.

### 6.5 Revision History

Date	Revision	Change
January 2008	1	Initial Release (Preliminary)
March 2008	2	Launch Release (Preliminary), Conversion to ON Semiconductor
January 2009	3	Updated assembly information and carrier details
July 2009	4	Updated part number

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